









Yassine Kali*, C. Fallaha, J. Rodas, M.Saad and F. Lesme Electrical Engineering Department École de Technologie Supérieure, Montreal, Canada E-mail: yassine.kali.1@ens.etsmtl.ca



Agenda



Context, Motivation and Contributions

02 – PRELIMINARIES

Quad-Rotor Model and Problem Formulation

03 – PROPOSED CONTROLLER

Concept and Design

04 – NUMERICAL SIMULATIONS

Results and Comparative Study

05 - CONCLUSIONS

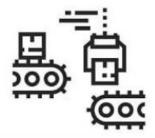
Summary and Ideas

1st Revolution



Mechanization, water and steam power

2nd Revolution



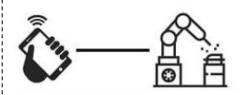
Mass production, assemply line and electrical energy

3rd Revolution



Electronics, computers and automation

4th Revolution



Cyber-physical systems and big-data

Late 18th century

Early 20th century

Early 70s

Today

Advantages

Vertical take-off and landing

Flying in low speed

High maneuverability



Small 4-rotors, less kinetic energy

Simple design and easy maintenance

Several possible applications

Disadvantages

underactuated system

Presence of hard nonlinearities

highly coupled dynamics



Requires precise & accurate rotorspeed control

It is sensitive to perturbation







Technique	Advantages	Drawbacks
Feedback Linearization	 Full linearization of the drone model Application of linear control methods 	 Holds higher order Lie derivatives States become noisy Parametric changes not handled Requires precise model
Integral Backstepping	Steady state error eliminationAugmented robustnessFinite-time convergence	Complexity in coefficient selectionIncrease the control effort
Conventional SMC	 Insensitive to external disturbances Robustness against model uncertainties Controller structure is simple and easily tunable Removes steady state error by adding integral action 	Chattering phenomenonSensor drift.



High Order SMC	 Reduces the chattering while preserving the SMC invariance property Insensitive to matched uncertainties Asymptotic convergence 	 Sliding surface selection is tedious Difficult implementation
Model Predictive Control (MPC)	 Optimal control input Desired state and input constraints can be defined Ability to predict future control moves 	 Additional computational power ne eded and storage required Hard to obtain a reliable prediction model which can lead to instability No robustness is ensured
Artificial Neural Networks (ANN) based control	 Learning ability Rejects disturbances and estimates uncertain model parameters No need of exact model Can be trained to provide tolerance against the cyber threats, injected faulty data, wireless communication attacks 	 Offline learning may fail under unknown environment Learning process is clumsy Requires larger computational effort due to stochastic learning policies

Technique	Advantages	Drawbacks
Feedback Linearization	 Full linearization of the drone model Application of linear control methods 	 Holds higher order Lie derivatives States become noisy Parametric changes not handled Requires precise model
Integral Backstepping	Steady state error eliminationAugmented robustnessFinite-time convergence	Complexity in coefficient selectionIncrease the control effort
Conventional SMC	 Insensitive to external disturbances Robustness against model uncertainties Controller structure is simple and easily tunable Removes steady state error by adding integral action 	Chattering phenomenonSensor drift



Original Contributions:

Exponential Power Reaching Law (EPRL) SMC with model-based sliding surfaces.

- <u>Simplicity of the control law:</u> A complete control decoupling is ensured since the proposed approach is not a model-based one.
- <u>Chattering reduction and finite-time reaching convergence</u>: The decoupled controller ensures chattering decoulping. In addition, the combination of the ERL and the PRRL ensures a faster finite-time convergence of the model-based sliding surfaces to zero and a chattering reduction since the switching gains are adapted according to the sliding surfaces' values.
- <u>Better tracking performance</u>: An augmented integral term in the proposed modelbased sliding surfaces helps removing the steady state error and helps rejecting the effects of unmatched uncertainties that acts on the systems' states.

System description

Quad-rotor UAV model has 6-DOF: i) $P = [x, y, z]^T$ is the vector of position (x, y) and altitude (z) states and ii) $\Theta = [\phi, \theta, \psi]^T$ is the vector of attitude or orientation or Euler angles (roll, pitch and yaw).

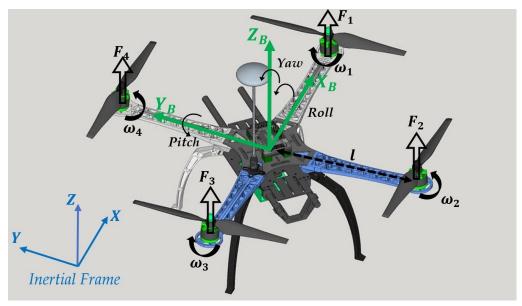


Figure 1: quad-rotor structure, forces, angles and frames



System description

Take off:

To take off the ground, the drone needs a net upward force. The motors generate lift that is greater than the force of gravity, making the drone take off.

Hovering:

Here, the motors create lift that is equal to the force of gravity on the system. Therefore, the lift and force of gravity cancel out and makes the drone hover in mid-air.

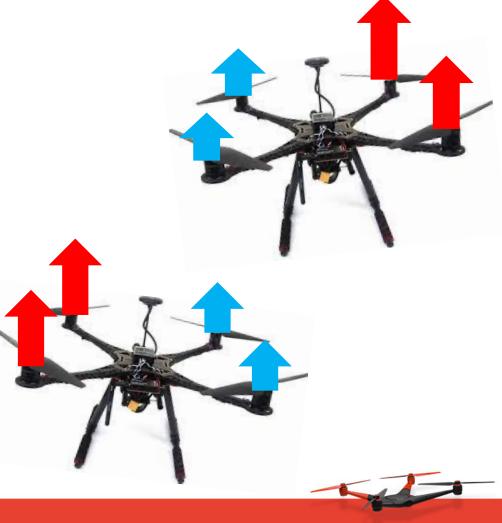




System description

Roll:

To Roll to the left, the lift is increased on the motors on the right. The drone must also decrease the lift on the motors on the left. Otherwise, to roll to the right, the drone must do the exact opposite.



System description

Pitch:

To make the drone pitch forwards (move towards) to you. The power applied to the rear motors is increased. This generates a forward net force which makes the drone's nose to pitch downward. The drone also have to decrease the power applied to the two front motors to keep the angular momentum conserved. The exact opposite is done to make the drone pitch backwards (move away) from you.

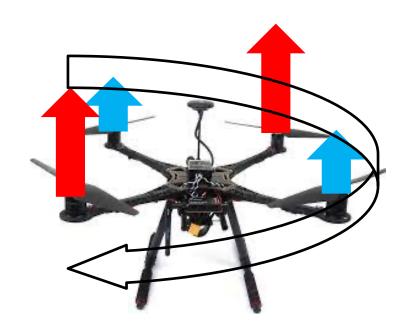




System description

Yaw:

To make the drone Yaw (rotate) clockwise. The drone must increase the lift on the anti-clockwise moving motors and also have to decrease the lift on clockwise rotating motors. The reason behind this is to keep the upward and downward net force equal to zero. There is also a resulting anti-clockwise torque. The drone rotates clockwise to conserve the angular momentum.





Dynamic model

Position and altitude model:

$$M\ddot{P} + K_P\dot{P} = \tau_P$$

- *M* : quad-rotor's mass;
- K_P : diagonal matrix where the elements are the drag coefficients of translation;
- τ_P : virtual input vector that is linked to the total thrust by the following formula:

$$\tau_T = \sqrt{\tau_{P1}^2 + \tau_{P2}^2 + (Mg + \tau_{P3})^2}$$

Attitude model:

$$I(\Theta)\ddot{\Theta} + C(\Theta, \dot{\Theta})\dot{\Theta} = \tau_{\Theta}$$

- $I(\Theta)$: inertia matrix;
- $C(\Theta, \dot{\Theta})$: Coriolis forces matrix;
- τ_{Θ} : vector of roll, pitch and yaw torques.



Problem formulation

Let $\tilde{P} = P - P^d \in \mathbb{R}^3$ be the position and altitude error vector with $P^d \in R^3$ is the desired known position and altitude vector and let $\tilde{\Theta} = \Theta - \Theta^d \in \mathbb{R}^3$ be the attitude error vector with $\Theta^d \in \mathbb{R}^3$ is the known desired attitude vector.

Assuming that the roll ϕ and the pitch θ are different from $\pm \pi/2$ to avoid singular configuration, the control objective is to design two cascade robust nonlinear controllers using EPRL SM fligh control with model-based sliding surfaces to ensure the convergence of both \tilde{P} and $\tilde{\Theta}$ to zero.



Global structure

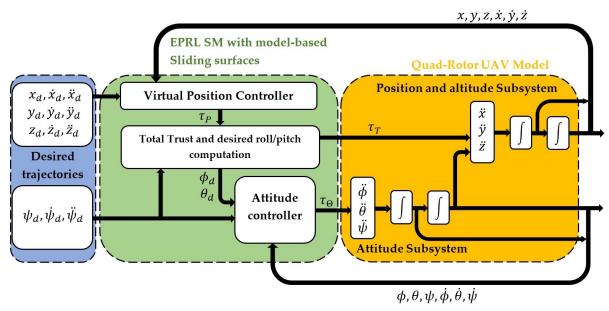


Figure 2: Closed-loop block diagram of the proposed flight controller



Outer control loop

The proposed approach design consists of two steps:

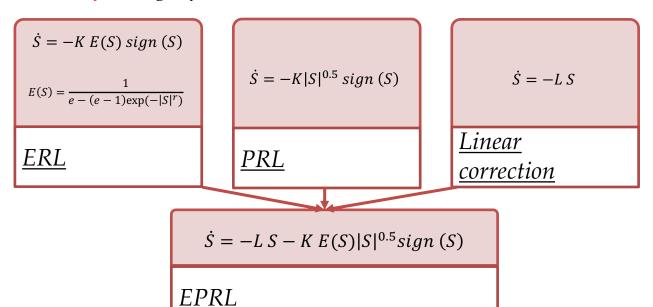
First Step: Design of the model-based sliding surfaces and compute its first-time derivative:

$$S_P = M\dot{\tilde{P}} + (K_P + \Lambda_1)\tilde{P} + \Lambda_2 \int_0^t \tilde{P} dt$$
$$\dot{S}_P = M(\ddot{P} - \ddot{P}^d) + (K_P + \Lambda_1)(\dot{P} - \dot{P}^d) + \Lambda_2 \tilde{P}$$



Outer control loop

Second step: Design of the control law:





Outer control loop

Theorem 1:

Consider the position model of the quad-rotor UAV system, the computed total force:

$$\tau_T = \sqrt{\tau_{P1}^2 + \tau_{P2}^2 + (Mg + \tau_{P3})^2}$$

$$\tau_{p} = \left[\tau_{p1}, \tau_{p2}, \tau_{p3}\right]^{T} = M\ddot{P}^{d} + K_{d}\dot{P}^{d} - \Lambda_{1}\dot{\tilde{P}} - \Lambda_{2}\tilde{P} - L_{P}S_{P} - K_{P}E_{P}(S_{P})|S_{P}|^{0.5}sign\left(S_{P}\right)$$

ensures the convergence of S_{Pi} to zero in a finite-time smaller than:

$$T_{Pi}^c \le \frac{-2}{L_{Pi}} \ln \left(\frac{K_{Pi} + L_{Pi} |S_{Pi}(t_0)|^{0.5}}{K_{Pi}} \right)$$



Inner control loop

Let us first compute the desired roll and pitch trajectories:

$$\phi^{d} = \arcsin\left(\frac{\sin(\psi^{d})\tau_{P1} - \cos(\psi^{d})\tau_{P2}}{\tau_{T}}\right)$$

$$\theta^{d} = \arctan\left(\frac{\cos(\psi^{d})\tau_{P1} + \sin(\psi^{d})\tau_{P2}}{\tau_{P3} + gM}\right)$$



Inner control loop

First Step: Design of the model-based sliding surfaces and compute its first-time derivative:

$$S_{\theta} = I(\theta)\dot{\theta} - I(\theta^{d})\dot{\theta}^{d} + \Gamma_{1}\tilde{\theta} + \Gamma_{2}\int_{0}^{t} \left(\tilde{\theta} - \left(W(\theta,\dot{\theta}) + C(\theta,\dot{\theta})\right)\dot{\theta} + \left(W(\theta^{d},\dot{\theta}^{d}) + C(\theta^{d},\dot{\theta}^{d})\right)\dot{\theta}^{d}\right)dt$$

$$W(\theta,\dot{\theta}) = \dot{I}(\theta) - 2C(\theta,\dot{\theta}), \qquad W(\theta^{d},\dot{\theta}^{d}) = \dot{I}(\theta^{d}) - 2C(\theta^{d},\dot{\theta}^{d})$$

$$\dot{S}_{\theta} = I(\theta)\ddot{\theta} + C(\theta,\dot{\theta})\dot{\theta} - \left(\underbrace{I(\theta^{d})\ddot{\theta}^{d} + C(\theta^{d},\dot{\theta}^{d})\dot{\theta}^{d}}_{\tau_{\theta}^{d}}\right) + \Gamma_{1}\dot{\tilde{\theta}} + \Gamma_{2}\tilde{\theta}$$

Second step: Compute the control law:

$$\dot{S}_{\Theta} = -L_{\Theta} S_{\Theta} - K_{\Theta} E_{\Theta}(S_{\Theta}) |S_{\Theta}|^{0.5} sign (S_{\Theta})$$



Inner control loop

Theorem 2:

Consider the attitude model of the quad-rotor system, the computed control input vector:

$$\tau_{\Theta} = \tau_{\Theta}^{d} - \Gamma_{1}\dot{\tilde{\Theta}} - \Gamma_{2}\tilde{\Theta} - L_{\Theta}S_{\Theta} - K_{\Theta}E_{\Theta}(S_{\Theta})|S_{\Theta}|^{0.5}sign(S_{\Theta})$$

ensures the convergence of S_{Pi} to zero in a finite-time smaller than:

$$T_{\Theta i}^{c} \leq \frac{-2}{L_{\Theta i}} \ln \left(\frac{K_{\Theta i} + L_{\Theta i} |S_{\Theta i}(t_0)|^{0.5}}{K_{\Theta i}} \right)$$



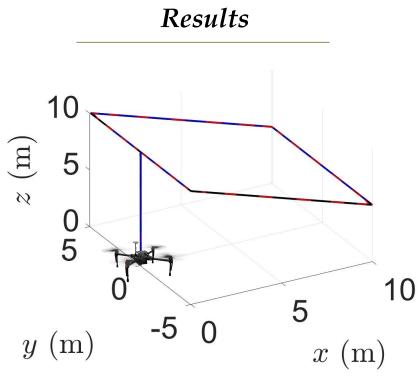


Figure 3: 3D cartesian space tracking

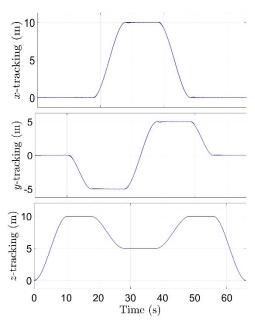


Figure 4: Position and altitude tracking

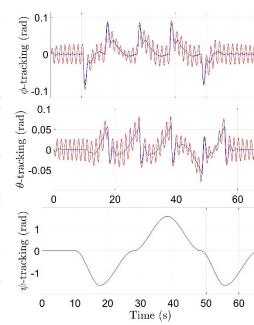
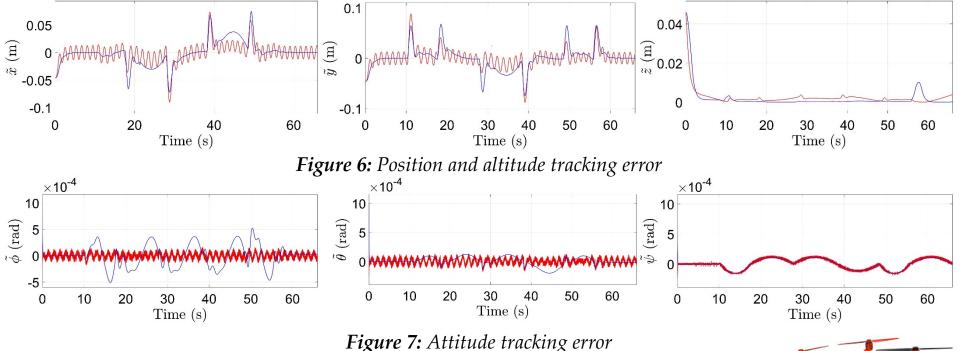


Figure 5: Attitude tracking



Results



Results

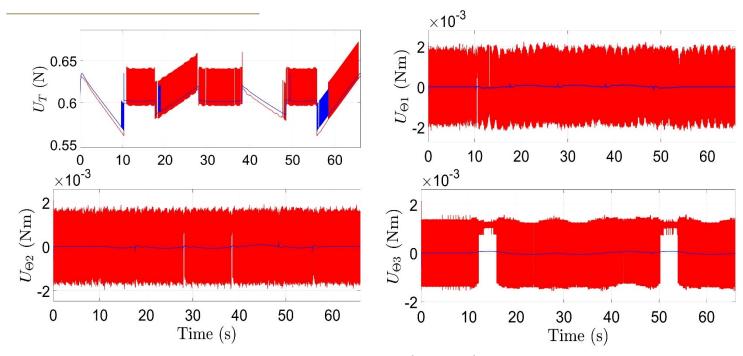
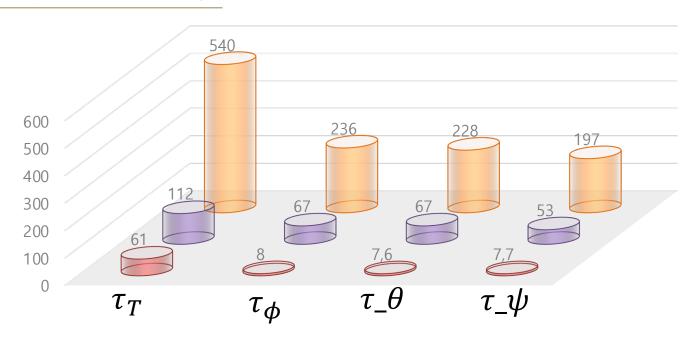


Figure 8: Computed control inputs



Comparative study



■ Proposed approach ■ Conventional SM with model-based SF ■ SMC

05 – Conclusions

Summary

- Development of EPRL SM flight control with model-based switching surfaces for quad-rotor UAV systems;
- Simulation results and quantitative comparison of the enhanced method with the conventional SM with model-based switching surfaces.

Future works

• Real-time implementation of the proposed approach on a real quad-rotor system and on other second-order nonlinear systems.





Hope you like this presentation

